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Teaching Adaptive Filters and Applications in Electrical and Computer Engineering Technology Program

Abstract

In this paper, we present our pedagogy and our experiences with teaching adaptive filters combined with applications in an advanced digital signal processing (DSP) course. This course is the second DSP course offered in the electrical and computer engineering technology (ECET) program according to the current trend of the DSP industry and students’ interests in their career development. A significant component of this course is adaptive filtering applications. A prerequisite for students is a working knowledge of the Laplace transform, Fourier series, Fourier transform, z-transform, discrete Fourier transform, digital filter design, and real-time DSP coding skills with TMS320C6713, a high-performance floating-point digital signal processor (DSP board) by Texas Instruments, acquired through the first DSP course. Although adaptive filtering is an exciting topic which allows for the exploration of many real-life applications, teaching this topic is often challenging due to its relatively heavy reliance on advanced mathematics. It is possible for the traditional mathematics used for the adaptive filtering theory to be minimized so that engineering technology students can more easily understand and grasp key concepts. With the MATLAB software tool, students can simulate and verify different adaptive filtering applications. To enhance hands-on learning, students are required to implement adaptive filtering techniques taught during the lectures using a TMS320C6713. Furthermore, it can be shown that a TMS320C6713 with its stereo channels proves an effective and flexible tool for various DSP implementations.

In this paper, we first focus on describing the pedagogy for teaching adaptive filter principles along with MATLAB simulations; then, we illustrate real-time DSP hands-on labs and projects with various applications. We will examine the assessment based on our collected data from course evaluations, student surveys and course work, and finally we will address possible improvement based on our assessment.

I. Introduction

The application of adaptive digital filtering technology has been found widely in modern electronic products, communication and control systems, computer peripherals, and multimedia devices. In the area of digital signal processing (DSP) education for the engineering technology curriculum, the adaptive filter theory and techniques are considered to be advanced topics to be covered during the student’s senior year. The trend of using adaptive filters in the industry has generated an increasing demand for engineering technology graduates with this particular working knowledge and skills. Many engineering technology programs have already offered a standard initial DSP course in their undergraduate curricula, particularly in the electrical and computer engineering technology (ECET) curriculum. The first DSP course generally covers basic techniques such as finite impulse response (FIR) filter design, infinite impulse response (IIR) filter design, FIR and IIR filter applications, discrete Fourier transform (DFT), and the signal spectral algorithm. To keep the pace with the DSP industry, our ECET curriculum has advanced to offer a second DSP course covering a broad range of topics which
include adaptive filtering, waveform compression and coding, multi-rate DSP, image and video processing.

Teaching adaptive filters to ECET students in the second DSP course appears challenging because it requires applying relatively advanced mathematics with the optimization theory in a matrix form to develop concepts. However, with our teaching pedagogy, this barrier can be overcome through illustrating the topic using a single coefficient adaptive filter. MATLAB is a necessary tool used to verify adaptive theory and perform simulations of various adaptive filtering applications. To motivate our technology students oriented about hands-on experience, we required them to perform real-time DSP using a floating-point digital signal processor \cite{8,9}, TMS320C6713 (development starter kit), to develop a comprehensive adaptive filtering project such as noise cancellation, and to demonstrate their working projects in class.

In this paper, we will describe the course prerequisites, course topics, and outline learning outcomes. With a focus on adaptive filtering techniques, we will describe our teaching pedagogy, MATLAB simulations, and hands-on real-time DSP labs and projects. Finally, we will examine the course assessment according to our collected data from the course evaluations, student surveys and course work, and then we will address possible improvement based on our assessment.

II. Learning Outcomes and Laboratories

The adaptive filter techniques are covered in our advanced DSP course (ECET 499) offered during the senior year, involving a 16-week class schedule with three-hour lectures and three-hour labs each week. Four weeks are allocated to covering adaptive filters. Prior to this second DSP course, the pre-requisite courses are: Introduction to Microcontrollers (ECET 209), Circuit Analysis Courses (ECET 207, ECET 257), Analog Network Signal Processing (ECET 307), and real-time DSP (ECET 357). Figure 1 shows a flowchart for the related courses.

![Flowchart of DSP-related courses.](image)

As shown in Figure 1, students in ECET209 gain background knowledge about the processor architecture, interface concepts, and basic C programming skills. The required skills of complex algebra and circuit analysis are covered in circuit courses such as AC Circuit Analysis (ECET 207) and Power and RF Electronics (ECET 257). Our Analog Network Signal Processing (ECET 307) course \cite{10} studies very important fundamental subjects: the Laplace transform, circuit
analysis using the Laplace transform, the Fourier series, and filter design concepts. These subjects are used extensively in the real-time DSP course (ECET 357), in which FIR filter design, IIR filter design, discrete Fourier transform (DFT), and signal spectrum and their applications are investigated. More importantly, the students are equipped with programming skills for using a floating-point digital signal processor, TMS320C6713, after successfully completing the first DSP course (ECET 357).

The advanced DSP course (ECET 499) as shown in Figure 1 covers the following key topics: (1) adaptive filters with applications such as noise cancellation, system modeling and echo cancellation; (2) waveform coding and compression including pulse code modulation (PCM), mu-law compression, adaptive differential PCM (ADPCM) and windowed discrete-cosine transform (DCT); (3) multi-rate DSP including sampling rate conversion and polyphase implementations; (4) image equalization and noise filtering; (5) image segmentation, pseudo-color generation and JPEG data compression; (6) other advanced DSP applications. These topics are covered through Chapters 10 to 13 in the DSP textbook. The labs and projects are listed in Table 1.

<table>
<thead>
<tr>
<th>Table 1. List of labs for ECET 499.</th>
</tr>
</thead>
<tbody>
<tr>
<td>Lab 1. Adaptive noise cancelling</td>
</tr>
<tr>
<td>Lab 2. Adaptive system modeling</td>
</tr>
<tr>
<td>Lab 3. PCM codec, mu-law compression, and ADPCM codec, transform coding with applications to speech signal</td>
</tr>
<tr>
<td>Lab 4. Sampling rate conversion and polyphase implementations</td>
</tr>
<tr>
<td>Lab 5. Image processing basics</td>
</tr>
<tr>
<td>Lab 6. Image processing: edge detection, pseudo color generation and JPEG color image compression</td>
</tr>
<tr>
<td>Project: Real-time DSP project: tonal noise cancellation</td>
</tr>
</tbody>
</table>

Notice that for labs 1-4 and course projects, students are required to perform MATLAB simulations first and then are required to focus on hands-on real-time DSP implementations using the TMS320C6713 board(s). The specific learning outcomes for adaptive filtering techniques are listed below:

Learning outcome 1: Given an objective function such as the mean squared error (MSE) function, develop an adaptive filter that seeks to minimize the given objective function by iteratively adjusting its parameters (such as its impulse response coefficients) to achieve the design goal in real time and under varying conditions.

Learning outcome 2: Apply real-time adaptive filtering techniques for applications such as noise cancellation and system modeling.

Labs 1-2 and course project listed in Table 1 are designed to fulfill these learning outcomes.

A. Pedagogy for Teaching Adaptive Filter Techniques

Our pedagogy for teaching adaptive filters includes the following steps: (1) using a single coefficient FIR filter to develop its Wiener filter solution; (2) introducing a single coefficient
steepest descent algorithm; (3) developing a single coefficient least mean square (LMS) algorithm; (4) extending a single coefficient LMS adaptive FIR filter to a standard LMS adaptive FIR filter; (5) performing MATLAB simulations for concept verifications.

Figure 2 shows a Wiener filter for noise cancellation, where a single coefficient filter is used for illustration; that is, \( y(n) = w x(n) \). \( w \) is the adaptive coefficient and \( y(n) \) is the Wiener filter output, which approximates the noise \( n(n) \) in the corrupted signal.

\[
d(n) = s(n) + n(n)
\]

\[
\text{Wiener filter}
\]

\[
y(n)
\]

\[
\text{Output}
\]

\[
e(n)
\]

Figure 2 Wiener filter for noise cancellation.

The enhanced signal \( e(n) \) is given by

\[
e(n) = d(n) - w x(n)
\]

Taking statistical expectation of the square of error leads to a quadratic function

\[
J = \sigma^2 - 2wP + w^2R
\]

where \( J = E[e^2(n)] \) is the mean square error or output power. When it is minimized, the noise power is maximally reduced. Since \( \sigma^2 = E[d^2(n)] \) (auto-correlation or power of the corrupted signal), \( P = E[d(n)x(n)] \) (cross-correlation), and \( R = E[x^2(n)] \) are constants, \( J \) is a quadratic function of \( w \) as shown below:

The best coefficient (optimal) \( w^* \) is unique which is corresponding to the minimum MSE error \( J_{\text{min}} \). By taking derivative of \( J \) and setting it to zero leads the solution as

\[
w^* = R^{-1}P
\]

A couple of numerical examples for finding the minimum locations of quadratic functions are given to develop the optimization concepts. Next, a steepest descent algorithm that is capable of minimizing the MSE function sample by sample to locate the filter coefficient(s) is introduced as follows:
\[ w_{n+1} = w_n - \mu \frac{dJ}{dw} \]

\( \mu = \) constant controlling speed of convergence and \( \frac{dJ}{dw} \) is the gradient of the MSE function.

The steepest descent method is more efficient since the matrix inverse of \( R^{-1} \) (for \( N \) filter coefficients) is not needed. Several numerical examples with a single coefficient filter are provided to show how iterations approach the optimal coefficient. To develop the LMS algorithm in terms of the sample based processing, a gradient \( \frac{dJ}{dw} \) can be approximated by its instantaneous value; that is,

\[ \frac{dJ}{dw} \approx 2(d(n) - wx(n)) \frac{d(d(n) - wx(n))}{dw} = -2e(n)x(n) \]

Substituting the instantaneous \( \frac{dJ}{dw} \) to the steepest descent algorithm, the LMS algorithm for updating a single coefficient is achieved:

\[ w_{n+1} = w_n + 2\mu e(n)x(n) \]

Finally by omitting iteration time index \( n \) and extending one coefficient filter to \( N \) coefficient filter, a standard LMS algorithm is obtained and listed in Table 2.

**Table 2. LMS adaptive FIR filter with \( N \) filter coefficients.**

<table>
<thead>
<tr>
<th>Step</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>Initialize ( w(0), w(1), \ldots, w(N-1) ) to arbitrary values</td>
</tr>
<tr>
<td>2</td>
<td>Read ( d(n), x(n) ), and perform digital filtering</td>
</tr>
<tr>
<td></td>
<td>( y(n) = w(0)x(n) + w(1)x(n-1) + \cdots + w(N-1)x(n-N+1) )</td>
</tr>
<tr>
<td>3</td>
<td>Compute the output error</td>
</tr>
<tr>
<td></td>
<td>( e(n) = d(n) - y(n) )</td>
</tr>
<tr>
<td>4</td>
<td>Update each filter coefficient using the LMS algorithm</td>
</tr>
<tr>
<td></td>
<td>for ( i = 0, \ldots, N-1 )</td>
</tr>
<tr>
<td></td>
<td>( w(i) = w(i) + 2\mu e(n)x(n-i) )</td>
</tr>
</tbody>
</table>

To illustrate the functionality of an adaptive filter, MATLAB examples for noise cancellation and system modeling are provided. The principle of noise cancellation is shown in Figure 4.

![Figure 4](image-url)  
**Figure 4.** Noise canceller using an adaptive filter.

The noise cancellation system in Figure 4 is assumed to have the following specifications: a sample rate = 8000 Hz; speech corrupted by a Gaussian noise with a power of 1 and delayed by 5 samples from a white noise source; and an adaptive FIR filter used to remove the noise, where
the number of FIR filter coefficients = 21 and the convergence factor for the LMS algorithm is chosen to be 0.01. The speech waveforms and speech spectral plots for the original, corrupted reference noise and the clean one are plotted in Figure 5, respectively. It is observed that the enhanced speech waveform and spectrum are very close to the originals. The LMS algorithm converges after approximately 400 iterations and is very effective for noise cancellation.

Figure 5. Original speech, corrupted speech, reference noise and clean speech.

Another typical MATLAB example is system modeling, in which an adaptive FIR filter is applied to track the behavior of an unknown system by monitoring the unknown system’s input and output. Figure 6 depicts the configuration of the system modeling.

Figure 6 Adaptive filter for system modeling.
As shown in the figure, \( y(n) \) is approaching the unknown system output during the adaptive process. Since both the unknown system and the adaptive filter use the same input, the transfer function of the adaptive filter will approximate the unknown system’s transfer function. In the simulation, the unknown system shown in Figure 6 is assumed to be a 4\(^{th}\)-order band-pass IIR filter, in which 3-dB lower and upper cut-off frequencies are 1400 Hz and 1600 Hz with a sampling rate of 8000 Hz. The system input contains 500 Hz, 1500 Hz, and 2500 Hz tones and its waveform \( x(n) \) is displayed in the left top plot in Figure 7. The output of the unknown system is expected to contain the 1500 Hz tone only, since the other two tones are rejected by the unknown system. An adaptive FIR filter with 21 coefficients is used to model the band-pass unknown system, and the convergence factor of the adaptive filter is set to be 0.01. In the time domain, the output waveforms of the unknown system \( d(n) \) and adaptive filter output \( y(n) \) are almost identical after the LMS algorithm converges around 70 samples. The error signal \( e(n) \) is also plotted to show how the adaptive filter keeps tracking the unknown system output with no difference.

![Figure 7](image_url)

Figure 7 Unknown system output, adaptive filter output and error output.

In the frequency domain, the first plot shows the input frequency components of 500 Hz, 1500 Hz, and 2500 Hz. The second plot shows the unknown system output spectrum with the 1500 Hz tone while the third plot displays the adaptive filter output spectrum with the 1500 Hz tone. It is apparent that the adaptive filter tracks the characteristics of the unknown system.

### B. Real-time Laboratory Content

Each adaptive filtering lab consists of two portions: MATLAB simulation and real-time DSP implementation with a single DSP board. The MATLAB simulation must be completed prior to the real-time implementation. We only describe real-time hands-on labs and projects. A single DSP board setup and program segment for verifying input and output signals are shown in Figure 8a and Figure 8b, respectively, where the sampling rate is 8000 samples per second.
Figure 8a Real-time DSP laboratory setup.

```c
float xL[1]={0.0};
float xR[1]={0};
float yL[1]={0.0};
float yR[1]={0,0};
interrupt void c_int11()
{
    float lc; /*left channel input */
    float rc; /*right channel input */
    float lcnew; /*left channel output */
    float rcnew; /*right channel output */
    int i;
    //Left channel and right channel inputs
    AIC23_data.combo=input_sample();
    lc=(float) (AIC23_data.channel[LEFT]);
    rc=(float) (AIC23_data.channel[RIGHT]);
    // Insert DSP algorithm below
    xL[0]=lc; /* Input from the left channel */
    xR[0]=rc; /* Input from the right channel */
    yL[0]=xL[0]; /* simplest DSP equation for the left channel*/
    yR[0]=xR[0]; /* simplest DSP equation for the right channel*/
    // End of the DSP algorithm
    lcnew=yL[0];
    rcnew=yR[0];
    AIC23_data.channel[LEFT]=(short) lcnew;
    AIC23_data.channel[RIGHT]=(short) rcnew;
    output_sample(AIC23_data.combo);
}
```
A configuration for Lab 1 (adaptive noise cancellation) is shown in Figure 9a, where the primary signal is a generated sinusoid and corrupted internally by a reference noise from the analog to digital conversion (ADC) channel. The reference noise can be set as a tonal noise fed via a function generator or from other noise source using microphone and amplifier. The output (difference between the corrupted signal and the adaptive filter output) presents a restored clean signal, similar to the original one from the internal digital oscillator. The sample program segment is given in Figure 9b.

```c
float b[2]={0.0, 0.587785}; /* Numerical coefficients for the 800 Hz digital oscillator */
float a[3]={1, -1.618034, 1}; /* Denominator coefficients for the digital oscillator */
float x[2]={5000, 0.0}; /* Set up the input as an impulse function */
float yy[3]={0.0, 0.0, 0.0};
float x[n][20]={0, 0, 0, 0, 0}; /* Reference input buffer */
float w[20]={0, 0, 0, 0, 0, 0, 0, 0, 0, 0, 0}; /* adaptive filter coefficients */
float d[1]={0.0}; /* Corrupted signal */
float y[1]={0.0}; /* Adaptive filter output */
float e[1]={0.0}; /* Enhanced signal */
float mu=0.0000000004; /* Adaptive filter convergence factor */
interrupt void c_int11()
{
    float lc; /* left channel input */
    float rc; /* right channel input */
    float lcnew; /* left channel output */
    float rcnew; /* right channel output */
    int i;
    // Left channel and right channel inputs
    AIC23_data.combo=input_sample();
    lc=(float) (AIC23_data.channel[LEFT]);
    rc=(float) (AIC23_data.channel[RIGHT]);
    /* Adaptive filter implementation */
    for (i=0; i<20; i++)
    {
        d[i]=y[i]+mu*e[i]*x[i];
        y[i]=d[i];
    }
    // Output processing
    e[0]=y[0]-rc;
    // Add other processing here...
    // Plotting or further analysis...
}
```

Figure 8b. Program segment for verifying input and output.

Figure 9a. Laboratory setup for noise cancellation using a LMS adaptive filter.
The implementation of system modeling in lab 2 is displayed in Figure 10a, where the input is fed from a function generator. The unknown system is a band-pass filter with a lower cut-off frequency of 1400 Hz and an upper cut-off frequency of 1600 Hz. When the input frequency is swept from 200 Hz to 3000 Hz, the output shows a maximum peak when the frequency is dialed to around 1500 Hz. Hence, the adaptive filter acts like the unknown system. Figure 10b shows the sample program segment.
Figure 10a. Laboratory setup for system modeling using a LMS adaptive filter.

/*Numerical coefficients */
/*for the bandpass filter (unknown system) with fL=1.4 kHz, fH=1.6 kHz*/
float b[5] = { 0.005542761540433, 0.000000000000002, -0.011085523080870, 
0.000000000000003, 0.005542761540431 }; 
/*Denominator coefficients */
/*for the bandpass filter (unknown system) with fL=1.4 kHz, fH=1.6 kHz*/
float a[5] = { 1.000000000000000, -1.450496619180500, 2.306093105231476, 
-1.297577189144526, 0.800817049322883 }; 
float xn[40] = {0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0 }; 
/*Reference input buffer*/
float w[40] = {0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0}; 
/*adaptive filter coefficients*/
float d[1] = {0.0}; /*Unknown system output */
float y[1] = {0.0}; /* Adaptive filter output */
float e[1] = {0.0}; /* Error signal */
float mu = 0.00000000002; /*Adaptive filter convergence factor*/

interrupt void c_int11()
{
    float lc; /*left channel input */
    float rc; /*right channel input */
    float lcnew; /*left channel output */
    float rcnew; /*right channel output */
    int i;
    //Left channel and right channel inputs
    AIC23_data.combo = input_sample();
    lc = (float) (AIC23_data.channel[LEFT]);
    rc = (float) (AIC23_data.channel[RIGHT]);
    // Insert DSP algorithm below
    for (i = 39; i > 0; i --) /*Update the input buffer*/
        { x[i] = x[i-1]; } 
    x[0] = lc;
    // Adaptive filter
    y[0] = 0;
    for (i = 0; i < 40; i++)
        { y[0] = y[0] + w[i]*x[i]; } 
    e[0] = d[0] - y[0]; /* Error output */
    for (i = 0; i < 40; i++)
        { w[i] = w[i] + 2*mu*e[0]*x[i]; } /* LMS algorithm */
    // End of the DSP algorithm
    lcnew = y[0]; /* Send the tracked output */
    rcnew = e[0]; /* Send the error signal*/
    AIC23_data.channel[LEFT] = (short) lcnew;
    AIC23_data.channel[RIGHT] = (short) rcnew;
    output_sample(AIC23_data.combo);
}

Figure 10b. Program segment for system modeling in Lab 2.
Finally, a course project is assigned. Students are divided into groups (two persons per group) to develop their selected DSP projects and to generate their design reports. At the end of the semester, everybody is required to demonstrate his/her project to the entire class so that the students can learn from each other. During the project’s developing phase, students can obtain advice from their instructor and work on their projects under the supervision of their instructor.

Figure 11 shows an example of a tonal noise reduction system that uses two TI DSP boards. The first DSP board is used to create a real-time corrupted signal mixed by the mono audio source (Left Line In [LCI1]) from any audio device, and the tonal noise (Right Line In [RCI1]) generated from the function generator. The output (Left line Out [LCO1]) is the corrupted signal, which is fed to the second DSP board for a noise cancellation application. The second DSP board receives the corrupted signal \( d(n) = s(n) + n(n) \) from channel LCI2 (Left Line In). The tonal noise \( n(n) \) in the corrupted signal is then cancelled by the adaptive FIR filter output \( y(n) \) using the reference noise input \( x(n) \) (Right Line In [RCI2]). Finally, the enhanced signal \( e(n) \) is sent to the output (Left Line Out [LCO2]) to produce a clean mono audio signal.

(a) Block diagram for adaptive noise cancellation

(b) Laboratory setup for adaptive noise cancellation.

Figure 11 Tonal noise cancellation with the LMS adaptive FIR filter.
III. Learning Outcome Assessment

The assessment presented here consists of a total of 12 student responses and is derived from our collected data from the course offered in spring of 2010 and fall of 2010. At the end of each semester, we conducted a student self-assessment. A student survey was given before the final exam asking each student to evaluate his/her achievement in the learning outcomes. Students were asked to select one of the following five (5) choices: understand well, understand, somewhat understand, somewhat confused, and confused. For statistical purposes, the five choices were assigned the scores of 5, 4, 3, 2, and 1, respectively. The average rating scores for learning outcomes are listed in row 2 Table 3.

Table 3. Student survey and instructor assessment.

<table>
<thead>
<tr>
<th>Learning Outcomes</th>
<th>O1</th>
<th>O2</th>
</tr>
</thead>
<tbody>
<tr>
<td>Student Survey</td>
<td>4.3</td>
<td>4.5</td>
</tr>
<tr>
<td>Instructor assessment</td>
<td>4.1</td>
<td>4.6</td>
</tr>
</tbody>
</table>

An instructor assessment based on the final exams and laboratory projects was also conducted. In order to do this, we designed a final exam in which the course learning outcome 1 was covered by the problems. We then computed the average points from all the students for the problem(s). The average rating on a scale from 1 to 5 was obtained by dividing the average points by the designated points for that problem(s) and then multiplying the result by 5. Outcome 2 was similarly assessed based on the student’s adaptive filter labs and projects. The instructor assessment is included in row 3 of Table 3.

The rating scores from the student survey and the ones from the instructor were consistent. The rating for course learning outcome 1 (O1) had a slightly bigger gap, in which the score from instructor rating was lower than the one from the student survey by 0.2. There is a smaller gap for O2. These discrepancies indicate that our ECET students were strong in hand-on applications.

We also conducted another student survey regarding our real-time adaptive filter labs. A set of questions, which are listed in Table 4, was given to students for evaluation. The students were allowed to select one of the following five (5) choices: strongly agree, agree, somewhat agree, disagree and strongly disagree. The corresponding rating scores were designated as 5, 4, 3, 2, and 1, respectively. Each average rating score is listed in Table 5.

Table 4. Survey questions for real-time DSP labs.

| Q1. Do you feel you can significantly grasp concepts of adaptive signal processing by working on real-time DSP coding? |
| Q2. Are you excited to work on real-time adaptive filter labs and projects? |
| Q3. Do you feel that real-time adaptive filtering is absolutely necessary in this course? |
| Q4. Does real-time coding of adaptive filters improve your problem solving ability? |
| Q5. Do you want more real-time adaptive filtering labs and projects in this course? |
| Q6. Do you want to learn more on DSP subjects and applications if possible in |
Table 5. Student survey for real-time adaptive filter labs.

<table>
<thead>
<tr>
<th>Question No.</th>
<th>Q1</th>
<th>Q2</th>
<th>Q3</th>
<th>Q4</th>
<th>Q5</th>
<th>Q6</th>
</tr>
</thead>
<tbody>
<tr>
<td>Student Survey</td>
<td>4.3</td>
<td>5.0</td>
<td>4.8</td>
<td>4.7</td>
<td>4.8</td>
<td>4.8</td>
</tr>
</tbody>
</table>

The average rating scores shown in Table 5 indicated the following:

1. Students strongly agreed that learning adaptive filters should involve real-time adaptive filter labs, which improved their problem solving ability.
2. Most of the students remained excited about the course, since the hands-on real-time laboratories motivated them.
3. Students were eager to learn more about real-time adaptive filter subjects.
4. The rating score for grasping adaptive concepts from adaptive filter labs was relatively low in comparison to the others. This indicated that students had not only learnt concepts from labs but also from class lectures, homework assignments, and MATLAB simulations.

V. Course Improvement

Based on our experiences in teaching adaptive filters with real-time DSP labs, we felt that the topic is well covered with suitable lectures and laboratories. We also felt that one particular prerequisite course, Real-time Digital Signal Processing (ECET 357), plays a very important role for students’ success in the second DSP course. We will continue to keep the standard in developing student real-time DSP skills. We would also like to encourage students to develop more comprehensive and challenging projects in the area of adaptive filtering such as echo cancellation, active noise control, adaptive IIR filters, etc.

VI. Conclusion

We have presented our pedagogy and our experiences with teaching adaptive filter techniques with both MATLAB simulations and real-time adaptive filter laboratories. From our assessment, we have found that hands-on real-time adaptive filter labs with practical applications can motivate students to achieve learning objectives and can increase students’ levels of interest in the area of DSP. We also expect that students will apply their gained knowledge and skills in adaptive filters to their senior design projects.

Bibliography